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Abstract. We study time-harmonic Maxwell's equations in meta-materials that use either perfect conductors or high-contrast materials. Based on known effective equations for perfectly conducting inclusions, we calculate the transmission and reflection coefficients for four different geometries. For high-contrast materials and essentially two-dimensional geometries, we analyze parallel electric and parallel magnetic fields and discuss their potential to exhibit transmission through a sample of meta-material. For a numerical study, one often needs a method that is adapted to heterogeneous media; we consider here a Heterogeneous Multiscale Method for high contrast materials. The qualitative transmission properties, as predicted by the analysis, are confirmed with numerical experiments. The numerical results also underline the applicability of the multiscale method.

Key words. homogenization, Maxwell's equations, multiscale method, meta-material

AMS subject classifications. 35B27, 35Q61, 65N30, 78M40

1 Introduction

Motivation. We study the transmission and reflection properties of meta-materials, i.e., of periodic microstructures of a composite material with two components. The interest in meta-materials has immensely grown in the last years as they exhibit astonishing properties such as band gaps or negative refraction; see [22, 34, 29]. The propagation of electromagnetic waves in such materials is modelled by time-harmonic Maxwell's equations for the electric field E and the magnetic field H :

$$\begin{cases} \operatorname{curl} E = i\omega\mu_0\mu H, & (1.1a) \\ \operatorname{curl} H = -i\omega\varepsilon_0\varepsilon E. & (1.1b) \end{cases}$$

We use the standard formulation with $\mu_0, \varepsilon_0 > 0$ the permeability and permittivity of vacuum, μ and ε the corresponding relative parameters, and $\omega > 0$ the imposed frequency. While most materials are non-magnetic, i.e., $\mu = 1$, the electric permittivity ε covers a wide range. In this paper, we study meta-materials consisting of air (i.e., $\varepsilon = 1$) and a (metal) microstructure Σ_η . The microstructure is assumed to be an η -periodic repetition of scaled copies of some geometry

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Σ . In the present study, we investigate in detail four different geometries: Σ can be a metal cylinder (in two rotations), a metal plate, or the complement of an air cylinder; see Fig. 2.2 and (2.6)–(2.9) for a detailed definition. For the electric permittivity in the microstructure Σ_η , we consider two different cases: perfect conductors that are formally obtained by setting $\varepsilon = \infty$, and high-contrast materials with $\varepsilon = \varepsilon_1 \eta^{-2}$, where $\varepsilon_1 \in \mathbb{C}$ is some complex number with $\text{Im}(\varepsilon_1) > 0$. In both cases, our study is based on the effective equations for the electric and magnetic field in the limit $\eta \rightarrow 0$.

The numerical simulation of electromagnetic wave propagation in such meta-materials is very challenging because of the rapid variations in the electric permittivity. Standard methods require the resolution of the η -scale, which often becomes infeasible even with today’s computational resources. Instead, we resort to homogenization and multiscale methods to extract macroscopic features and the behaviour of the solution. The effective equations obtained by homogenization can serve as a good motivation and starting point in this process.

Literature. Effective equations for Maxwell’s equations in meta-materials are obtained in several different settings with various backgrounds in mind: Dielectric bulk inclusions with high-contrast media [6, 7, 13] can explain the effect of artificial magnetism and lead to unusual effective permeabilities μ , while long wires [5] lead to unusual effective permittivities. A combination of both structures is used to obtain a negative-index meta-material in [27]. Topological changes in the material in the limit $\eta \rightarrow 0$, such as found in split rings [10], also incite unusual effective behaviour. Perfect conductors were recently studied as well: split rings in [28] and different geometries in [36]. Finally, we briefly mention that the Helmholtz equation—as the two-dimensional reduction of Maxwell’s equations—is often studied as the first example for unusual effective properties: high-contrast inclusions in [8] or high-contrast layer materials in [11], just to name a few. An overview on this vast topic is provided in [35].

Concerning the numerical treatment, we focus on the Heterogeneous Multiscale Method (HMM) [19, 20]. For the HMM, first analytical results concerning the approximation properties for elliptic problems have been derived in [1, 21, 32] and then extended to other problems, such as time-harmonic Maxwell’s equations [25] and the Helmholtz equation and Maxwell’s equations with high-contrast [33, 37]. Another related work is the multiscale asymptotic expansion for Maxwell’s equations [12]. For further recent contributions to HMM approximations for Maxwell’s equations we refer to [16, 26]. Sparse tensor product finite elements for multiscale Maxwell-type equations are analyzed in [14] and an adaptive generalized multiscale finite element method is studied in [15].

Main results. We perform an analytical and a numerical study of transmission properties of meta-materials that contain either perfect conductors or high-contrast materials. The main results are the following:

- 1.) Using the effective equations of [36], we calculate the reflection and transmission coefficients for four microscopic geometries Σ . Few geometrical parameters are sufficient to fully describe the effective coefficients. We show that only certain polarizations can lead to transmission.
- 2.) For the two geometries that are invariant in the e_3 -direction, we study the limit behaviour of the electromagnetic fields for high-contrast media. When the electric field is parallel to e_3 , all fields vanish in the limit. Instead, when the magnetic field is parallel to e_3 , transmission cannot be excluded due to resonances.
- 3.) Extensive numerical experiments for high-contrast media confirm the analytical results. The numerical experiments underline the applicability of the Heterogeneous Multiscale Method to these challenging setting.

Some further remarks on 2.) are in order. The results are related to homogenization results of [7, 13], but we study more general geometries, since the highly conducting material can be connected. Furthermore, the results are related to [8, 11], where connected structures are inves-

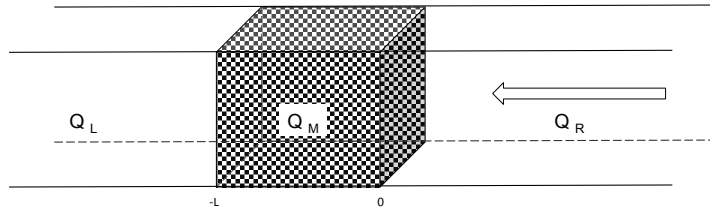


Figure 2.1: Waveguide domain G with periodic scatterer Σ_η contained in the middle part Q_M and incident wave from the right.

tigated, but in a two-dimensional formulation. We treat here properties of the three-dimensional solutions. We emphasize that the transmission properties of a high-contrast medium cannot be captured in the framework of perfect conductors, since the latter excludes resonances on the scale of the periodicity (except if three different length-scales are considered as in [28]).

Organization of the paper. The paper is organized as follows: In Section 2, we detail the underlying problem formulations and revisit existing effective equations. In Section 3, we compute the transmission coefficients for perfect conductors and derive effective equations for high-contrast media. In Section 4, we briefly introduce the Heterogeneous Multiscale Method. Finally, in Section 5 we present several numerical experiments concerning the transmission properties of our geometries for high-contrast materials.

2 Problem formulation and effective equations

This section contains the precise formulation of the problem, including the description of the four microscopic geometries. We summarize the relevant known homogenization results and apply them to the cases of interest.

2.1 Geometry and material parameters

We study the time-harmonic Maxwell equations with linear material laws. The geometry is periodic with period $\eta > 0$; solutions depend on this parameter and are therefore indexed with η . On a domain $G \subset \mathbb{R}^3$, the problem is to find $E^\eta, H^\eta : G \rightarrow \mathbb{C}^3$, such that

$$\begin{cases} \operatorname{curl} E^\eta = i\omega\mu_0 H^\eta, & (2.1a) \\ \operatorname{curl} H^\eta = -i\omega\varepsilon_0\varepsilon_\eta E^\eta, & (2.1b) \end{cases}$$

subject to appropriate boundary conditions. In the following, we will give details on the geometry G and on the choice of the material parameter ε_η , the relative permittivity. Note that the system allows to eliminate one unknown. Indeed, if we insert H^η from (2.1a) into (2.1b), we obtain

$$\operatorname{curl} \operatorname{curl} E^\eta = \omega^2 \mu_0 \varepsilon_0 \varepsilon_\eta E^\eta. \quad (2.2)$$

Alternatively, substituting E^η from (2.1b) into (2.1a), we obtain

$$\operatorname{curl} \varepsilon_\eta^{-1} \operatorname{curl} H^\eta = \omega^2 \mu_0 \varepsilon_0 H^\eta. \quad (2.3)$$

Geometry. As sketched in Fig. 2.1, with positive numbers $\ell_2, \ell_3 > 0$, the unbounded macroscopic domain is the waveguide domain

$$G := \{x = (x_1, x_2, x_3) \in \mathbb{R}^3 : x_2 \in (-\ell_2, \ell_2) \text{ and } x_3 \in (-\ell_3, \ell_3)\}. \quad (2.4)$$

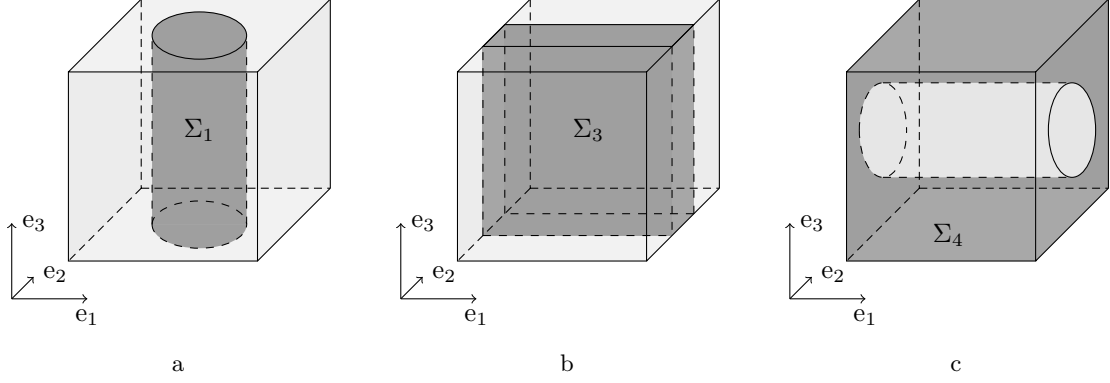


Figure 2.2: The cube shows the periodicity cell Y . The microstructures Σ_1 , Σ_3 , and Σ_4 are shown in dark grey. (a) The metal cylinder Σ_1 . (b) The metal plate Σ_3 . (c) The metal part Σ_4 is the complement of a cylinder.

With another positive number $L > 0$, the domain is divided into three parts (left, middle, right) as

$$Q_L := \{x \in G : x_1 \leq -L\}, \quad Q_M := \{x \in G : x_1 \in (-L, 0)\}, \quad \text{and} \quad Q_R := \{x \in G : x_1 \geq 0\}.$$

The scatterer Σ_η is contained in the middle part Q_M . For the boundary conditions, we consider an incident wave from the right that travels along the x_1 -axis to the left. We restrict ourselves here to normal incidence. For the analysis, we impose periodic boundary conditions on the lateral boundaries of the domain G . For the numerics, we will modify the boundary conditions slightly: we truncate G in x_1 -direction (to obtain a bounded domain) and consider impedance boundary conditions (with the incident wave as data) on the whole boundary of G .

The scatterer Σ_η is given as an η -periodic structure. We use the periodicity cell $Y := [-\frac{1}{2}, \frac{1}{2}]^3$ and introduce the set I_η of all vectors such that a scaled and shifted copy of Y is contained in Q_M , $I_\eta := \{j \in \mathbb{Z}^3 \mid \eta(j + Y) \subset Q_M\}$. A set $\Sigma \subset Y$ specifies the meta-material, which is defined as

$$\Sigma_\eta := \bigcup_{j \in I_\eta} \eta(j + \Sigma). \quad (2.5)$$

For the microscopic structure Σ we consider the following four examples. The metal cylinder (see Fig. 2.2a) is defined for $r \in (0, 1/2)$ as

$$\Sigma_1 := \{y = (y_1, y_2, y_3) \in Y : y_1^2 + y_2^2 < r^2\}. \quad (2.6)$$

The set Σ_2 is obtained by a rotation which aligns the cylinder with the e_1 -axis,

$$\Sigma_2 := \{y = (y_1, y_2, y_3) \in Y : y_2^2 + y_3^2 < r^2\}. \quad (2.7)$$

To define the metal plate (see Fig. 2.2b), we fix $r \in (0, 1/2)$ and set

$$\Sigma_3 := \{y = (y_1, y_2, y_3) \in Y : y_2 \in (-r, r)\}. \quad (2.8)$$

The fourth geometry is obtained by removing an ‘‘air cylinder’’ from the unit cube (see Fig. 2.2c); for $r \in (0, 1/2)$ we set

$$\Sigma_4 := Y \setminus \{y = (y_1, y_2, y_3) \in Y : y_2^2 + y_3^2 < r^2\}. \quad (2.9)$$

Material parameters. We recall that all materials are non-magnetic, the relative magnetic permeability is $\mu \equiv 1$. Outside the central region, there is no scatterer; we hence set $\varepsilon_\eta = 1$ in Q_L and Q_R . The middle part Q_M contains Σ_η . We set $\varepsilon_\eta = 1$ in $Q_M \setminus \Sigma_\eta$. It remains to specify the electric permittivity ε_η in Σ_η . We consider two different settings.

(PC) In the case of *perfect conductors*, we set, loosely speaking, $\varepsilon_\eta = +\infty$ in Σ_η . More precisely, we require that E^η and H^η satisfy (2.1) in $G \setminus \overline{\Sigma}_\eta$ and $E^\eta = H^\eta = 0$ in Σ_η . Boundary conditions are induced on $\partial\Sigma_\eta$: The magnetic field H^η has a vanishing normal component and the electric field E^η has vanishing tangential components on $\partial\Sigma_\eta$.

(HC) In the case of *high-contrast media*, we define the permittivity as

$$\varepsilon_\eta(x) := \begin{cases} \frac{\varepsilon_1}{\eta^2} & \text{if } x \in \Sigma_\eta, \\ 1 & \text{if } x \in G \setminus \Sigma_\eta, \end{cases} \quad (2.10)$$

where $\varepsilon_1 \in \mathbb{C}$ with $\operatorname{Re}(\varepsilon_1) > 0$, $\operatorname{Im}(\varepsilon_1) > 0$. Physically speaking, this means that the scatterer Q_M consists of periodically disposed metal inclusions Σ_η embedded in vacuum. The scaling with η^2 means that the optical thickness of the inclusions remains constant; see [7].

In both settings and throughout this paper, we consider sequences of solutions $(E^\eta, H^\eta)_\eta$ to (2.1) which are bounded in $L^2(G; \mathbb{C}^3)$,

$$\sup_{\eta > 0} \int_G |E^\eta|^2 + |H^\eta|^2 < \infty. \quad (2.11)$$

Let us remark that the specific geometry of the microstructures Σ_1, Σ_2 , and Σ_4 is not important; the cylinders could as well be cuboids.

2.2 Effective equations

Homogenization theory allows to consider the limit $\eta \rightarrow 0$. One identifies limiting fields \hat{E} and \hat{H} (the latter does not coincide with the weak limit of H^η) and limiting equations for these fields. Using the tool of two-scale convergence, such results have been obtained for perfect conductors as well as for high-contrast materials. We briefly summarize the main findings here; analysis and numerics below are built upon these results.

Perfect conductors (PC). The homogenization analysis for this case has been performed in [36]. Since the parameters of vacuum are used outside the scatterer, the original Maxwell equations describe the limiting fields in Q_L and Q_R . In the meta-material Q_M , however, different equations hold. There holds $E^\eta \rightharpoonup \hat{E}$ and $H^\eta \rightharpoonup \hat{\mu}\hat{H}$ in $L^2(G)$ and the fields \hat{E} and \hat{H} solve

$$\begin{cases} \operatorname{curl} \hat{E} = i\omega\mu_0\hat{\mu}\hat{H} & \text{in } G, & (2.12a) \\ \operatorname{curl} \hat{H} = -i\omega\varepsilon_0\hat{\varepsilon}\hat{E} & \text{in } G \setminus Q_M, & (2.12b) \\ (\operatorname{curl} \hat{H})_k = -i\omega\varepsilon_0(\hat{\varepsilon}\hat{E})_k & \text{in } G, \text{ for every } k \in \mathcal{N}_\Sigma, & (2.12c) \\ \hat{E}_k = 0 & \text{in } Q_M, \text{ for every } k \in \mathcal{L}_\Sigma, & (2.12d) \\ \hat{H}_k = 0 & \text{in } Q_M, \text{ for every } k \in \mathcal{N}_{Y \setminus \overline{\Sigma}}. & (2.12e) \end{cases}$$

The effective coefficients $\hat{\mu}$ and $\hat{\varepsilon}$ are determined by cell-problems. For the cell-problems, details on the index sets, and the derivation of system (2.12), we refer to [36]. The index sets \mathcal{N}_Σ , \mathcal{L}_Σ , and $\mathcal{N}_{Y \setminus \overline{\Sigma}}$ are subsets of $\{1, 2, 3\}$ and can be determined easily from topological properties of Σ . Loosely speaking: An index k is in the set \mathcal{L}_Σ , if there is a curve (loop) that runs in Σ and

Table 2.1: Index sets \mathcal{N}_Σ , \mathcal{L}_Σ , and $\mathcal{N}_{Y \setminus \bar{\Sigma}}$ for microstructures Σ_1 to Σ_4 of (2.6)–(2.9).

geometry	metal cylinder Σ_1	metal cylinder Σ_2	metal plate Σ_3	air cylinder Σ_4
\mathcal{N}_Σ	$\{1, 2\}$	$\{2, 3\}$	$\{2\}$	\emptyset
\mathcal{L}_Σ	$\{3\}$	$\{1\}$	$\{1, 3\}$	$\{1, 2, 3\}$
$\mathcal{N}_{Y \setminus \bar{\Sigma}}$	\emptyset	\emptyset	$\{2\}$	$\{2, 3\}$

connects opposite faces of Y in direction e_k . An index k is in \mathcal{N}_Σ , if there is no loop of that kind. We collect the index sets \mathcal{N}_Σ , \mathcal{L}_Σ , and $\mathcal{N}_{Y \setminus \bar{\Sigma}}$ for the geometries Σ_1 to Σ_4 in Table 2.1.

We will specify equations (2.12c)–(2.12e) for the four chosen geometries in Section 3.1. With the effective equations for the perfect conductors at hand, one can ask for the transmission and reflection coefficients of the meta-material. This is the goal of our analysis in Section 3.1.

high-contrast media (HC). Homogenization results for high-contrast media are essentially restricted to the case of non-connected metal parts, i.e., to geometries that are obtained by Σ which is compactly embedded in Y (it does not touch the boundary of the cube); see, e.g., [6, 10, 7]. The few exceptions are mentioned below.

For such geometries, the limit equations have again the form of Maxwell’s equations,

$$\begin{cases} \operatorname{curl} \hat{E} = i\omega\mu_0\hat{\mu}\hat{H} & \text{in } G, \\ \operatorname{curl} \hat{H} = -i\omega\varepsilon_0\hat{\varepsilon}\hat{E} & \text{in } G. \end{cases} \quad (2.13a)$$

$$(2.13b)$$

In $Q_L \cup Q_R$, the effective fields coincide with the weak limits of the original fields, and the effective relative coefficients are unit tensors. In the meta-material Q_M , however, the high-contrast in the definition of the permittivity ε_η in (2.10) leads to non-trivial limit equations. The effective material parameters $\hat{\varepsilon}$ and $\hat{\mu}$ are obtained via cell problems and they can take values that are not to be expected from the choice of the material parameters in the η -problem.

As discussed in Section 2.1, time-harmonic Maxwell’s equations can equivalently be written as a single second order PDE for the H -field or the E -field. For the H -field we obtain

$$\operatorname{curl} \widehat{\varepsilon^{-1}} \operatorname{curl} \hat{H} = \omega^2 \varepsilon_0 \mu_0 \hat{\mu} \hat{H} \quad \text{in } G. \quad (2.14)$$

Again, the effective material parameters $\widehat{\varepsilon^{-1}}$ and $\hat{\mu}$ are defined via solutions of cell problems and we refer to [13, 37] for details. We remark that the equivalence of the two formulations (2.13) and (2.14) has been shown in [37]. In particular, the effective permeability $\hat{\mu}$ agrees between both formulations and we have the relation $\widehat{\varepsilon^{-1}} = (\hat{\varepsilon})^{-1}$.

The effective equations (2.13) or (2.14) mean that, in the limit $\eta \rightarrow 0$, the meta-material Q_M with high-contrast permittivity ε_η behaves like a homogeneous material with permittivity $\hat{\varepsilon}$ and permeability $\hat{\mu}$. The occurrence of a permeability $\hat{\mu}$ in the effective equations is striking and this effect is known as artificial magnetism; see [8]. Moreover, $\hat{\mu}$ depends on the frequency ω and it can have a negative real part for certain frequencies. Negative values of the permeability are caused by (Mie) resonances in the inclusions Σ and are studied in detail in [7, 37].

As mentioned, a crucial assumption for the homogenization analysis in [7, 13] is that Σ is compactly contained in the unit cube. For the four geometries Σ_1 to Σ_4 , this assumption is clearly not met; we therefore ask whether certain components of the effective fields \hat{E} and \hat{H} vanish in this case as in the case of perfect conductors. This motivates our analysis in Section 3.2 as well as the numerical experiments in Section 5.

Regarding known results on non-compactly contained inclusions we mention the thin wires in [9] and [27], and the dimensionally reduced analysis of the metal plates Σ_3 in [11].

3 Analysis of the microscopic geometries Σ_1 to Σ_4

In Section 3.1, we treat the case of perfect conductors and compute the transmission coefficients from the effective equations (2.12). In Section 3.2, we treat the case of high-contrast media and discuss the possibility of nontrivial transmission coefficients.

3.1 Transmission and reflection coefficients for perfect conductors

We compute the transmission and reflection coefficients for four different geometries: metal cylinders, metal plate, and air cylinder. We consider the waveguide $G = Q_L \cup Q_M \cup Q_R$ of Section 2.1 and impose periodic boundary conditions on the lateral boundary of G . We recall that the four microscopic structures Σ_1 to Σ_4 are defined in (2.6)–(2.9). Based on the effective equations (2.12) for the perfect conductors, we compute the transmission and reflection coefficients for these geometries.

Results for perfect conductors. Before we discuss the examples in detail, we present an overview of the results. The propagation of the electromagnetic wave in vacuum is described by the time-harmonic Maxwell equations

$$\begin{cases} \operatorname{curl} \hat{E} = i\omega\mu_0 \hat{H} & \text{in } Q_L \cup Q_R, \\ \operatorname{curl} \hat{H} = -i\omega\varepsilon_0 \hat{E} & \text{in } Q_L \cup Q_R. \end{cases} \quad (3.1a)$$

$$(3.1b)$$

For the electromagnetic fields, we use the time-convention $e^{-i\omega t}$. From (3.1) we deduce that both fields are divergence-free in $Q_L \cup Q_R$. We shall assume that the electric field $\hat{E}: G \rightarrow \mathbb{C}^3$ in Q_R is the superposition of a normalized incoming wave with normal incidence and a reflected wave:

$$\hat{E}(x) := (e^{-ik_0 x_1} + R e^{ik_0 x_1}) e_k, \quad (3.2)$$

for $x = (x_1, x_2, x_3) \in Q_R$ and $k \in \{2, 3\}$. Here, $R \in \mathbb{C}$ is the reflection coefficient and $k_0 = \omega\sqrt{\varepsilon_0\mu_0}$. Note that the electric field \hat{E} in (3.2) travels along the x_1 -axis from right to left.

Due to (3.1a), the effective magnetic field $\hat{H}: G \rightarrow \mathbb{C}^3$ is given by

$$\hat{H}(x) = (-1)^l \frac{k_0}{\omega\mu_0} (e^{-ik_0 x_1} - R e^{ik_0 x_1}) e_l, \quad (3.3)$$

where $l = 2$ if $k = 3$ and $l = 3$ if $k = 2$ and $x \in Q_R$. Equation (3.1b) is satisfied in Q_R by our choice of k_0 .

On the other hand, for the transmitted electromagnetic wave in the left domain Q_L , we make the ansatz

$$\hat{E}(x) = T e^{-ik_0(x_1+L)} e_k \quad \text{and} \quad \hat{H}(x) = (-1)^l \frac{k_0}{\omega\mu_0} T e^{-ik_0(x_1+L)} e_l, \quad (3.4)$$

where $T \in \mathbb{C}$ is the transmission coefficient. We recall that $L > 0$ is the width of the meta-material Q_M and $\{x_1 = -L\}$ is the interface between left and middle domain. Since the meta-material in Q_M can lead to reflections, the transmission coefficient $T \in \mathbb{C}$ does not necessarily satisfy $|T| = 1$; by energy conservation there always holds $|T| = 1 - |R|$.

Our results are collected in Table 3.1. The table lists transmission coefficients for the four geometries in the case that the incoming magnetic field H is parallel to e_3 .

In the remainder of this section we compute the transmission coefficient T and the reflection coefficient R for the four microscopic geometries and verify, in particular, the formulas of Table 3.1. Moreover, the effective fields in the meta-material Q_M are determined.

microstructure Σ	transmission coefficient T
metal cylinder Σ_1	$T = 4p_1\sqrt{\alpha\gamma}\left[(\alpha + \gamma)(1 - p_1^2) + 2\sqrt{\alpha\gamma}(1 + p_1^2)\right]^{-1}$
metal cylinder Σ_2	$T = 4p_2\sqrt{\gamma}\left[(1 + \gamma)(1 - p_2^2) + 2\sqrt{\gamma}(1 + p_2^2)\right]^{-1}$
metal plate Σ_3	$T = 4p_0\alpha\left[(1 + \alpha^2)(1 - p_0^2) + 2\alpha(1 + p_0^2)\right]^{-1}$
air cylinder Σ_4	$T = 0$

Table 3.1: Overview of the transmission coefficients T when H is parallel to e_3 . We see, in particular, that T is vanishing for the structure Σ_4 , but it is nonzero for the other micro-structures. The constant $\gamma \in \mathbb{C}$ depends on the microstructure and on solutions to cell problems, and is defined in the subsequent sections, $\alpha := |Y \setminus \Sigma|$ is the volume fraction of air, $L > 0$ is the width of the meta-material Q_M . We use $k_0 = \omega\sqrt{\varepsilon_0\mu_0}$ and the numbers $p_0 := e^{ik_0L}$, $p_1 := p_0 e^{i\sqrt{\alpha\gamma}L}$, and $p_2 := p_0 e^{i\sqrt{\gamma}L}$.

3.1.1 The metal cylinder Σ_1

The metal cylinder Σ_1 has a high symmetry, which allows to compute the effective permeability $\hat{\mu}$. To do so, we define the projection $\pi: Y \rightarrow \mathbb{R}^2$ onto the first two components, i.e., $\pi(y_1, y_2, y_3) := (y_1, y_2)$. Moreover, we set $Y^2 := \pi(Y)$ and $\Sigma_1^2 := \pi(\Sigma_1)$.

Choose $l \in \{1, 2\}$ and denote by $H^l \in L^2(Y; \mathbb{C}^3)$ the distributional periodic solution of

$$\begin{cases} \operatorname{curl} H^l = 0 \text{ in } Y \setminus \bar{\Sigma}_1, & (3.5a) \\ \operatorname{div} H^l = 0 \text{ in } Y, & (3.5b) \\ H^l = 0 \text{ in } \Sigma_1, & (3.5c) \end{cases}$$

with

$$\oint H^l = e_l. \quad (3.5d)$$

The normalization of the last equation is defined in [36]; loosely speaking, the left hand collects values of line integrals of H^l , where the lines are curves in $Y \setminus \Sigma$ and connect opposite faces of Y . Problem (3.5) is uniquely solvable by [36, Lemma 3.5]. Given the field $H^l = (H_1^l, H_2^l, H_3^l)$ we define the field $h^l: Y^2 \rightarrow \mathbb{C}^2$ as

$$h^l(y_1, y_2) := \int_0^1 (-H_2^l, H_1^l)(y_1, y_2, y_3) dy_3. \quad (3.6)$$

Lemma 3.1. *Let $H^l \in L^2(Y; \mathbb{C}^3)$ be the solution of (3.5). Then $h^l \in L^2(Y^2; \mathbb{C}^2)$ of (3.6) is a distributional periodic solution to the two-dimensional problem*

$$\begin{cases} \operatorname{div} h^l = 0 \text{ in } Y^2 \setminus \Sigma_1^2, & (3.7a) \\ \nabla^\perp \cdot h^l = 0 \text{ in } Y^2, & (3.7b) \\ h^l = 0 \text{ in } \Sigma_1^2. & (3.7c) \end{cases}$$

Moreover, there exists a potential $\psi \in H_{\#}^1(Y^2; \mathbb{C})$ such that $h^l = \nabla\psi - \delta_{2l}e_1 + \delta_{1l}e_2$.

Proof. The proof consists of a straightforward calculation. □

The decomposition of h^l allows to determine the effective permeability $\hat{\mu}$, which, by [36], is given as

$$\hat{\mu}(x) := \mu_{\text{eff}} \mathbf{1}_{Q_M}(x) + \text{Id} \mathbf{1}_{G \setminus \overline{Q_M}}(x), \quad (3.8)$$

where

$$(\mu_{\text{eff}})_{kl} := \int_Y H^l \cdot e_k. \quad (3.9)$$

Lemma 3.2 (Effective permeability for the metal cylinder). *For the microstructure $\Sigma = \Sigma_1$ the permeability μ_{eff} is given by*

$$\mu_{\text{eff}} = \text{diag}(1, 1, |Y \setminus \Sigma_1|). \quad (3.10)$$

Proof. To shorten the notation, we write $y' := (y_1, y_2) \in Y^2$. Applying Fubini's theorem and using the decomposition of h^1 , we find that

$$(\mu_{\text{eff}})_{11} = \int_Y H^1 \cdot e_1 = \int_{Y^2} h_2^1(y') \, dy' = \int_{Y^2} \partial_2 \psi(y') \, dy' + |Y^2| = 1,$$

where, in the last equality, we exploited that ψ is Y^2 -periodic and that $|Y^2| = 1$. A similar computation shows that $(\mu_{\text{eff}})_{22} = 1$.

To compute $(\mu_{\text{eff}})_{12}$, we note that $h_1^1(y') = \partial_1 \psi(y')$. Applying Fubini's theorem, we find

$$(\mu_{\text{eff}})_{12} = \int_Y H^1 \cdot e_2 = - \int_{Y^2} h_1^1(y') \, dy' = - \int_{Y^2} \partial_1 \psi(y') \, dy' = 0.$$

As $h_2^2(y') = \partial_2 \psi(y')$, we can proceed as before and find $(\mu_{\text{eff}})_{21} = 0$.

One readily checks that $H^3(y) := \mathbf{1}_{Y \setminus \Sigma_1}(y) e_3$ is the solution of the cell problem (3.5) with $\mathcal{f} H^3 = e_3$. The missing entries of the effective permeability matrix μ_{eff} can now be computed using the formula for H^3 and the definition of μ_{eff} in (3.9). \square

Besides $\hat{\mu}$, we also need the effective permittivity $\hat{\varepsilon}$. For $l \in \{1, 2, 3\}$ we denote by $E^l \in L^2(Y; \mathbb{C}^3)$ the weak periodic solution to

$$\begin{cases} \text{curl } E^l = 0 \text{ in } Y, & (3.11a) \\ \text{div } E^l = 0 \text{ in } Y \setminus \overline{\Sigma_1}, & (3.11b) \\ E^l = 0 \text{ in } \Sigma_1, & (3.11c) \end{cases}$$

with

$$\int_Y E^l = e_l. \quad (3.11d)$$

Problem (3.11) is uniquely solvable by [36, Lemma 3.1]. Consequently, the solutions to (3.11) are real vector fields. Indeed, for each index $l \in \{1, 2, 3\}$ the vector field $\text{Im}(E^l): Y \rightarrow \mathbb{R}^3$ is a weak solution to (3.11) with $\int_Y \text{Im}(E^l) = 0$ and hence $\text{Im}(E^l) = 0$ in Y .

As in [36] we set

$$\hat{\varepsilon}(x) := \varepsilon_{\text{eff}} \mathbf{1}_{Q_M}(x) + \text{Id} \mathbf{1}_{G \setminus \overline{Q_M}}(x), \quad (3.12)$$

where

$$(\varepsilon_{\text{eff}})_{kl} := \int_Y E^k \cdot E^l. \quad (3.13)$$

Lemma 3.3 (Effective permittivity for the metal cylinder). *For the microstructure $\Sigma = \Sigma_1$, the permittivity ε_{eff} is given by*

$$\varepsilon_{\text{eff}} = \text{diag}(\gamma, \gamma, 0), \quad (3.14)$$

where $\gamma := (\varepsilon_{\text{eff}})_{1,1}$.

Proof. As shown in Table 2.1, we find that $\mathcal{N}_{\Sigma_1} = \{1, 2\}$. From [36, Lemma 3.2] we hence deduce that $(\varepsilon_{\text{eff}})_{k,3}$ as well as $(\varepsilon_{\text{eff}})_{3,k}$ vanish for all $k \in \{1, 2, 3\}$. We claim that the matrix ε_{eff} is symmetric. Because of $(\varepsilon_{\text{eff}})_{k,3} = (\varepsilon_{\text{eff}})_{3,k} = 0$ we only have to prove that $(\varepsilon_{\text{eff}})_{1,2} = (\varepsilon_{\text{eff}})_{2,1}$. As the solutions E^1 and E^2 of the cell problem (3.11) are real vector fields, we compute that

$$(\varepsilon_{\text{eff}})_{1,2} = \int_Y E^1 \cdot E^2 = \int_Y E^2 \cdot E^1 = (\varepsilon_{\text{eff}})_{2,1}.$$

To show that $(\varepsilon_{\text{eff}})_{1,2} = (\varepsilon_{\text{eff}})_{2,1} = 0$, we consider the map $M: Y \rightarrow Y$ that is defined by the diagonal matrix $\text{diag}(-1, 1, 1)$. Note that $M(\Sigma_1) = \Sigma_1$. To shorten the notation, we set $E := E^1$. Consider the vector field $F: Y \rightarrow \mathbb{R}^3$,

$$F(x) := ME(Mx) = \begin{pmatrix} -E_1 \\ E_2 \\ E_3 \end{pmatrix} (-x_1, x_2, x_3).$$

One readily checks that F is a solution to the cell problem (3.11) with $\int_Y F = -e_1$. Due to the unique solvability of the cell problem (3.11), we conclude that that $F = -E$. Similarly, we find that $ME^2 \circ M = E^2$. Thus

$$(\varepsilon_{\text{eff}})_{1,2} = \int_Y E^1 \cdot E^2 = - \int_Y ME^1(My) \cdot ME^2(My) dy = - \int_Y E^1 \cdot E^2 = -(\varepsilon_{\text{eff}})_{1,2}.$$

Hence $(\varepsilon_{\text{eff}})_{1,2} = (\varepsilon_{\text{eff}})_{2,1} = 0$.

We are left to prove $(\varepsilon_{\text{eff}})_{2,2} = (\varepsilon_{\text{eff}})_{1,1}$. To do so, we consider the rotation map $R: Y \rightarrow Y$ which is defined by the matrix

$$\begin{pmatrix} 0 & -1 & 0 \\ 1 & 0 & 0 \\ 0 & 0 & 1 \end{pmatrix}.$$

Then $R(\Sigma_1) = \Sigma_1$. Moreover, as the cell problem (3.11) is uniquely solvable, we find that $RE^2 \circ R = -E^1$. Thus

$$(\varepsilon_{\text{eff}})_{1,1} = \int_Y E^1 \cdot E^1 = \int_Y RE^2(Ry) \cdot RE^2(Ry) dy = \int_Y E^2 \cdot E^2 = (\varepsilon_{\text{eff}})_{2,2}.$$

This proves the claim. \square

By Theorem 4.1 of [36], the microstructure Σ_1 together with the effective permittivity from (3.14) and permeability from (3.10) implies that the effective equations are

$$\begin{cases} \partial_2 \hat{H}_3 - \partial_3 \hat{H}_2 = -i\omega\varepsilon_0(\hat{\varepsilon}\hat{E})_1 & \text{in } G, & (3.15a) \end{cases}$$

$$\begin{cases} \partial_3 \hat{H}_1 - \partial_1 \hat{H}_3 = -i\omega\varepsilon_0(\hat{\varepsilon}\hat{E})_2 & \text{in } G, & (3.15b) \end{cases}$$

$$\begin{cases} \hat{E}_3 = 0 & \text{in } Q_M. & (3.15c) \end{cases}$$

The equations (3.15) do not repeat (2.12a) and (2.12b). Due to (2.12a), the effective electric field is divergence-free. As we assume that \hat{E} travels along the x_1 -axis, the first component \hat{E}_1 vanishes. Due to (3.15c) we expect no transmission if the effective electric field is polarized in e_3 -direction. For nontrivial transmission, we may therefore make the following ansatz for the effective electric field $\hat{E}: G \rightarrow \mathbb{C}^3$,

$$\hat{E}(x) := (e^{-ik_0x_1} + R e^{ik_0x_1}) e_2 \quad \text{for } x = (x_1, x_2, x_3) \in Q_R.$$

Thanks to (2.12b) the magnetic field \hat{H} is given by

$$\hat{H}(x) = -\frac{k_0}{\omega\mu_0} (e^{-ik_0x_1} - R e^{ik_0x_1}) e_3 \quad \text{for } x = (x_1, x_2, x_3) \in Q_R.$$

In the meta-material Q_M we write

$$\hat{E}(x) = (T_M e^{-ik_1x_1} + R_M e^{ik_1x_1}) e_2 \quad \text{and} \quad \hat{H}(x) = -\frac{k_1}{\omega\mu_0\alpha} (T_M e^{-ik_1x_1} - R_M e^{ik_1x_1}) e_3$$

for $x \in Q_M$, where we used equation (2.12a) and (3.10) to determine the magnetic field with $\alpha := |Y \setminus \Sigma_1|$. To compute the value of k_0 we use equation (2.12b) and we find that $k_0 = \omega\sqrt{\varepsilon_0\mu_0}$. From (3.15b) we deduce that $k_1 = k_0\sqrt{\alpha\gamma}$. In Q_L we choose (3.4) as the ansatz for \hat{E} and \hat{H} , where $k = 2$ and $l = 3$.

Lemma 3.4 (Transmission and reflection coefficients). *Given the electric and magnetic fields \hat{E} and \hat{H} as described above. Set $\alpha := |Y \setminus \Sigma_1|$, $k_1 = \omega\sqrt{\varepsilon_0\mu_0\alpha\gamma}$, and $p_1 := e^{ik_1L}$. The coefficients are then given by*

$$R = \frac{(\alpha - \gamma)(1 - p_1^2)}{(\alpha + \gamma)(1 - p_1^2) + 2\sqrt{\alpha\gamma}(1 + p_1^2)}, \quad T_M = \frac{2\sqrt{\alpha}(\sqrt{\alpha} + \sqrt{\gamma})}{(\alpha + \gamma)(1 - p_1^2) + 2\sqrt{\alpha\gamma}(1 + p_1^2)},$$

$$R_M = -\frac{2\sqrt{\alpha}p_1^2(\sqrt{\alpha} - \sqrt{\gamma})}{(\alpha + \gamma)(1 - p_1^2) + 2\sqrt{\alpha\gamma}(1 + p_1^2)}, \quad T = \frac{4\sqrt{\alpha\gamma}p_1}{(\alpha + \gamma)(1 - p_1^2) + 2\sqrt{\alpha\gamma}(1 + p_1^2)}.$$

Proof. By (2.12a) the tangential trace of \hat{E} has no jump across the surfaces $\{x \in G: x_1 = 0\}$ and $\{x \in G: x_1 = -L\}$. Thus

$$T_M + R_M = 1 + R \quad \text{and} \quad T = p_1 T_M + \frac{1}{p_1} R_M. \quad (3.16)$$

The effective field \hat{H} is parallel to e_3 and hence, by (3.15b), the third component \hat{H}_3 does not jump across the surfaces $\{x \in G: x_1 = 0\}$ and $\{x \in G: x_1 = -L\}$. We may therefore conclude that

$$\sqrt{\frac{\gamma}{\alpha}}(T_M - R_M) = 1 - R \quad \text{and} \quad T = \sqrt{\frac{\gamma}{\alpha}} \left(p_1 T_M - \frac{1}{p_1} R_M \right). \quad (3.17)$$

Here we used that $k_0 = \omega\sqrt{\mu_0\varepsilon_0}$ and $k_1 = k_0\sqrt{\alpha\gamma}$. Solving the equations on the left-hand side in (3.16) and (3.17) for R and the other two equations for T , we find that

$$T_M + R_M - 1 = R = 1 - \sqrt{\frac{\gamma}{\alpha}}(T_M - R_M) \quad (3.18)$$

and

$$p_1 T_M + \frac{1}{p_1} R_M = T = \sqrt{\frac{\gamma}{\alpha}} \left(p_1 T_M - \frac{1}{p_1} R_M \right). \quad (3.19)$$

Setting $d_+ := 1 + \sqrt{\gamma/\alpha}$ and $d_- := 1 - \sqrt{\gamma/\alpha}$, equations (3.18) and (3.19) can be written as

$$d_+ T_M = 2 - d_- R_M \quad \text{and} \quad p_1 d_- T_M = -\frac{1}{p_1} d_+ R_M. \quad (3.20)$$

Solving each of the two equations in (3.20) for R_M and then equating the two expressions for R_M , we obtain

$$T_M = \frac{2d_+}{d_+^2 - d_-^2 p_1^2} = \frac{2(1 + \sqrt{\gamma/\alpha})}{(1 + \sqrt{\gamma/\alpha})^2 - (1 - \sqrt{\gamma/\alpha})^2 p_1^2} = \frac{2\sqrt{\alpha}(\sqrt{\alpha} + \sqrt{\gamma})}{(\sqrt{\alpha} + \sqrt{\gamma})^2 - (\sqrt{\alpha} - \sqrt{\gamma})^2 p_1^2}.$$

Note that $(\sqrt{\alpha} + \sqrt{\gamma})^2 - (\sqrt{\alpha} - \sqrt{\gamma})^2 p_1^2 = (\alpha + \gamma)(1 - p_1^2) + 2\sqrt{\alpha\gamma}(1 + p_1^2)$, which yields the formula for T_M . From the second equation in (3.20), we deduce that

$$R_M = -p_1^2 \frac{d_-}{d_+} T_M = -p_1^2 \frac{\sqrt{\alpha} - \sqrt{\gamma}}{\sqrt{\alpha} + \sqrt{\gamma}} T_M = -\frac{2\sqrt{\alpha} p_1^2 (\sqrt{\alpha} - \sqrt{\gamma})}{(\alpha + \gamma)(1 - p_1^2) + 2\sqrt{\alpha\gamma}(1 + p_1^2)}.$$

By (3.18), we have that

$$\begin{aligned} R &= T_M + R_M - 1 = \frac{2\sqrt{\alpha}(\sqrt{\alpha} + \sqrt{\gamma}) - 2\sqrt{\alpha} p_1^2 (\sqrt{\alpha} - \sqrt{\gamma})}{(\alpha + \gamma)(1 - p_1^2) + 2\sqrt{\alpha\gamma}(1 + p_1^2)} - 1 \\ &= \frac{(\alpha - \gamma)(1 - p_1^2)}{(\alpha + \gamma)(1 - p_1^2) + 2\sqrt{\alpha\gamma}(1 + p_1^2)}. \end{aligned}$$

To compute the coefficient T we use equation (3.16) and find that

$$T = \frac{2\sqrt{\alpha}(\sqrt{\alpha} + \sqrt{\gamma})p_1 - 2\sqrt{\alpha}(\sqrt{\alpha} - \sqrt{\gamma})p_1}{(\alpha + \gamma)(1 - p_1^2) + 2\sqrt{\alpha\gamma}(1 + p_1^2)} = \frac{4\sqrt{\alpha}\sqrt{\gamma}p_1}{(\alpha + \gamma)(1 - p_1^2) + 2\sqrt{\alpha\gamma}(1 + p_1^2)}.$$

This proves the claim. \square

3.1.2 The metal cylinder Σ_2

Similar to the previous section, we shall determine the transmission and reflection coefficients for a metal cylinder, considering the microstructure Σ_2 . We define the effective permeability and the effective permittivity $\hat{\mu}, \hat{\varepsilon}: G \rightarrow \mathbb{C}^3$ as in (3.8) and (3.12). Following the reasoning of Section 3.1.1, we find that the μ_{eff} and ε_{eff} are given by

$$\mu_{\text{eff}} = \text{diag}(|Y \setminus \Sigma_2|, 1, 1) \quad \text{and} \quad \varepsilon_{\text{eff}} = \text{diag}(0, \gamma, \gamma),$$

where $\gamma \in \mathbb{C}$ is defined as $\gamma := \int_Y E^2 \cdot E^2$. The effective equations for the microstructure Σ_2 are

$$\begin{cases} \partial_3 \hat{H}_1 - \partial_1 \hat{H}_3 = -i\omega \varepsilon_0 (\hat{\varepsilon} \hat{E})_2 & \text{in } G, & (3.21a) \\ \partial_1 \hat{H}_2 - \partial_2 \hat{H}_1 = -i\omega \varepsilon_0 (\hat{\varepsilon} \hat{E})_3 & \text{in } G, & (3.21b) \\ \hat{E}_1 = 0 & \text{in } Q_M. & (3.21c) \end{cases}$$

We may take a similar ansatz for the effective fields as in Section 3.1.1 and obtain the following transmission and reflection coefficients. Note that k_1 in Section 3.1.1 has to be replaced by $k_2 := k_0 \sqrt{\gamma}$.

Lemma 3.5 (Transmission and reflection coefficients). *Within the setting of Section 3.1.1, we set $k_2 = \omega \sqrt{\varepsilon_0 \mu_0 \gamma}$ and $p_2 := e^{ik_2 L}$. The reflection and transmission coefficients for Σ_2 are given by*

$$\begin{aligned} R &= \frac{(1 - \gamma)(1 - p_2^2)}{(1 + \gamma)(1 - p_2^2) + 2\sqrt{\gamma}(1 + p_2^2)}, & T_M &= \frac{2(1 + \sqrt{\gamma})}{(1 + \gamma)(1 - p_2^2) + 2\sqrt{\gamma}(1 + p_2^2)}, \\ R_M &= -\frac{2p_2^2(1 - \sqrt{\gamma})}{(1 + \gamma)(1 - p_2^2) + 2\sqrt{\gamma}(1 + p_2^2)}, & T &= \frac{4p_2\sqrt{\gamma}}{(1 + \gamma)(1 - p_2^2) + 2\sqrt{\gamma}(1 + p_2^2)}. \end{aligned}$$

Note that in the above transmission and reflection coefficients the volume fraction of air $\alpha = |Y \setminus \Sigma_2|$ does not appear. This is different for the metal cylinder Σ_1 ; see Lemma 3.4.

Proof. Thanks to (2.12a) we know that the tangential components of \hat{E} do not jump across the surfaces $\{x \in G: x_1 = 0\}$ and $\{x \in G: x_1 = -L\}$. Hence

$$1 + R = T_M + R_M \quad \text{and} \quad T = p_2 T_M + \frac{1}{p_2} R_M. \quad (3.22)$$

The effective field \hat{H} is parallel to e_3 and hence, due to (3.21a), the third component \hat{H}_3 does neither jump across $\{x \in G: x_1 = 0\}$ nor across $\{x \in G: x_1 = -L\}$. We may therefore conclude that

$$1 - R = \sqrt{\gamma}(T_M - R_M) \quad \text{and} \quad T = \sqrt{\gamma} \left(p_2 T_M - \frac{1}{p_2} R_M \right). \quad (3.23)$$

Here we used that $k_2 = k_0 \sqrt{\gamma} = \omega \sqrt{\varepsilon_0 \mu_0 \gamma}$.

Solving the equations on the left-hand side in (3.22) and (3.23) for R and the other two for T , we find that

$$T_M + R_M - 1 = R = 1 - \sqrt{\gamma}(T_M - R_M) \quad \text{and} \quad p_2 T_M + \frac{1}{p_2} R_M = T = \sqrt{\gamma} \left(p_2 T_M - \frac{1}{p_2} R_M \right).$$

Setting $c_+ := 1 + \sqrt{\gamma}$ and $c_- := 1 - \sqrt{\gamma}$, these two equations can be re-written as

$$c_+ T_M = 2 - c_- R_M \quad \text{and} \quad c_- p_2 T_M = -\frac{1}{p_2} c_+ R_M. \quad (3.24)$$

We can solve for T_M and obtain

$$T_M = \frac{2c_+}{c_+^2 - c_-^2 p_2^2} = \frac{2(1 + \sqrt{\gamma})}{(1 + \sqrt{\gamma})^2 - (1 - \sqrt{\gamma})^2 p_2^2}.$$

Note that $(1 + \sqrt{\gamma})^2 - (1 - \sqrt{\gamma})^2 p_2^2 = (1 + \gamma)(1 - p_2^2) + 2\sqrt{\gamma}(1 + p_2^2)$, which proves the formula for T_M . By (3.24) we then conclude that

$$R_M = -p_2^2 \frac{c_-}{c_+} T_M = -p_2^2 \frac{1 - \sqrt{\gamma}}{1 + \sqrt{\gamma}} T_M = -\frac{2p_2^2(1 - \sqrt{\gamma})}{(1 + \gamma)(1 - p_2^2) + 2\sqrt{\gamma}(1 + p_2^2)}.$$

To determine the coefficient R we recall from above that $R = T_M + R_M - 1$ and hence

$$R = \frac{2(1 + \sqrt{\gamma}) - 2p_2^2(1 - \sqrt{\gamma}) - (1 + \sqrt{\gamma})^2 + (1 - \sqrt{\gamma})^2 p_2^2}{(1 + \gamma)(1 - p_2^2) + 2\sqrt{\gamma}(1 + p_2^2)} = \frac{(1 - \gamma)(1 - p_2^2)}{(1 + \gamma)(1 - p_2^2) + 2\sqrt{\gamma}(1 + p_2^2)}.$$

As $T = p_2 T_M + 1/p_2 R_M$, we find that

$$T = \frac{2p_2(1 + \sqrt{\gamma}) - 2p_2(1 - \sqrt{\gamma})}{(1 + \gamma)(1 - p_2^2) + 2\sqrt{\gamma}(1 + p_2^2)} = \frac{4p_2\sqrt{\gamma}}{(1 + \gamma)(1 - p_2^2) + 2\sqrt{\gamma}(1 + p_2^2)}.$$

This proves the claim. \square

We chose the same polarization for the electric and the magnetic field as in Section 3.1.1. By symmetry of the microstructure, we may as well assume that \hat{E} is parallel to e_3 and \hat{H} is parallel to e_2 and obtain the same reflection and transmission coefficients.

3.1.3 The metal plate

We consider the microstructure Σ_3 ; that is, a metal plate which is perpendicular to e_2 . Following the reasoning in Section 5.2 in [36], we determine the effective equations and obtain:

$$\begin{cases} \partial_3 \hat{H}_1 - \partial_1 \hat{H}_3 = -i\omega\varepsilon_0\alpha^{-1}\hat{E}_2 & \text{in } Q_M, & (3.25a) \\ \hat{E}_1 = \hat{E}_3 = 0 & \text{in } Q_M, & (3.25b) \\ \hat{H}_2 = 0 & \text{in } Q_M, & (3.25c) \end{cases}$$

where $\alpha := |Y \setminus \Sigma_3|$.

The electromagnetic wave is assumed to travel in e_1 -direction from right to left. Moreover, by (2.12a), the electric field is divergence free. Hence, the first component \hat{E}_1 vanishes. Because of (3.25b) we expect no transmission if the electric field is polarized in e_3 -direction. We may therefore make the following ansatz for the effective electric field $\hat{E}: G \rightarrow \mathbb{C}^3$,

$$\hat{E}(x) := (e^{-ik_0x_1} + R e^{ik_0x_1}) e_2 \quad \text{for } x = (x_1, x_2, x_3) \in Q_R.$$

Thanks to (2.12b), the magnetic field \hat{H} is given by

$$\hat{H}(x) = -\frac{k_0}{\omega\mu_0} (e^{-ik_0x_1} - R e^{ik_0x_1}) e_3 \quad \text{for } x \in Q_R.$$

By equation (3.25b), the first and the third component of the effective electric field are trivial; from this and equation (2.12a), we deduce that

$$\hat{E}(x) = (T_M e^{-ik_3x_1} + R_M e^{ik_3x_1}) e_2 \quad \text{and} \quad \hat{H}(x) = -\frac{k_3}{\omega\mu_0\alpha} (T_M e^{-ik_3x_1} - R_M e^{ik_3x_1}) e_3 \quad \text{in } Q_M.$$

The value of k_3 can be determined by (3.25a) and we find that $k_3 = k_0 = \omega\sqrt{\varepsilon_0\mu_0}$. In Q_L , we choose (3.4) as the ansatz for \hat{E} and \hat{H} , where $k = 2$ and $l = 3$.

Lemma 3.6 (Transmission and reflection coefficients). *Given the effective fields \hat{E} and \hat{H} as described above. Set $\alpha := |Y \setminus \Sigma_3|$ and $p_0 := e^{i\omega\sqrt{\varepsilon_0\mu_0}L}$. The reflection and transmission coefficients are given by*

$$R = \frac{(\alpha^2 - 1)(1 - p_0^2)}{(1 + \alpha^2)(1 - p_0^2) + 2\alpha(1 + p_0^2)}, \quad T_M = \frac{2\alpha(\alpha + 1)}{(1 + \alpha^2)(1 - p_0^2) + 2\alpha(1 + p_0^2)}, \quad (3.26)$$

$$R_M = -\frac{2\alpha p_0^2(\alpha - 1)}{(1 + \alpha^2)(1 - p_0^2) + 2\alpha(1 + p_0^2)}, \quad T = \frac{4p_0\alpha}{(1 + \alpha^2)(1 - p_0^2) + 2\alpha(1 + p_0^2)}. \quad (3.27)$$

Proof. From (2.12a) we deduce that $\text{curl } \hat{E}$ has no singular part and hence the tangential trace of \hat{E} along the surfaces $\{x \in G: x_1 = 0\}$ and $\{x \in G: x_1 = -L\}$ does not jump. Thus

$$T_M + R_M = 1 + R \quad \text{and} \quad T = p_0 T_M + \frac{1}{p_0} R_M. \quad (3.28)$$

As \hat{H} is parallel to e_3 , we deduce from (3.25a) that \hat{H}_3 does not jump across the surfaces $\{x \in G: x_1 = 0\}$ and $\{x \in G: x_1 = -L\}$. This implies that

$$1 - R = \frac{1}{\alpha} (T_M - R_M) \quad \text{and} \quad T = \frac{1}{\alpha} \left(p_0 T_M - \frac{1}{p_0} R_M \right). \quad (3.29)$$

Here we used that $k_0 = k_3 = \omega\sqrt{\varepsilon_0\mu_0}$. Note that $\alpha > 0$ and hence we find $a > 0$ such that $\sqrt{a} = 1/\alpha$. With this new parameter a , the equations in (3.29) read

$$\sqrt{a}(T_M - R_M) = 1 - R \quad \text{and} \quad T = \sqrt{a}\left(p_0 T_M - \frac{1}{p_0} R_M\right). \quad (3.30)$$

Thus the equations in (3.28) and (3.30) have the same structure as the equations in (3.22) and (3.23). We may therefore use the formulas for R, T, R_M , and T_M derived in Section 3.1.2. Note that

$$1 + \sqrt{a} = \frac{\alpha + 1}{\alpha} \quad 1 - \sqrt{a} = \frac{\alpha - 1}{\alpha}, \quad \text{and} \quad 1 - a = \frac{\alpha^2 - 1}{\alpha^2}.$$

Thus

$$\begin{aligned} T_M &= \frac{2(1 + \sqrt{a})}{(1 + \sqrt{a})^2 - (1 - \sqrt{a})^2 p_0^2} = \frac{2\alpha(\alpha + 1)}{(\alpha + 1)^2 - (\alpha - 1)^2 p_0^2} = \frac{2\alpha(\alpha + 1)}{(1 + \alpha^2)(1 - p_0^2) + 2\alpha(1 + p_0^2)}, \\ R_M &= -\frac{2p_0^2(1 - \sqrt{a})}{(1 + \sqrt{a}) - (1 - \sqrt{a})p_0^2} = -\frac{2\alpha p_0^2(\alpha - 1)}{(\alpha + 1)^2 - (\alpha - 1)^2 p_0^2} = -\frac{2\alpha p_0^2(\alpha - 1)}{(1 + \alpha^2)(1 - p_0^2) + 2\alpha(1 + p_0^2)}, \\ R &= \frac{(1 - a)(1 - p_0^2)}{(1 + \sqrt{a}) - (1 - \sqrt{a})p_0^2} = \frac{(\alpha^2 - 1)(1 - p_0^2)}{(\alpha + 1)^2 - (\alpha - 1)^2 p_0^2} = \frac{(\alpha^2 - 1)(1 - p_0^2)}{(1 + \alpha^2)(1 - p_0^2) + 2\alpha(1 + p_0^2)}, \end{aligned}$$

and

$$T = \frac{4p_0\sqrt{a}}{(1 + \sqrt{a})^2 - (1 - \sqrt{a})^2 p_0^2} = \frac{4p_0\alpha}{(\alpha + 1)^2 - (\alpha - 1)^2 p_0^2} = \frac{4p_0\alpha}{(1 + \alpha^2)(1 - p_0^2) + 2\alpha(1 + p_0^2)}.$$

This proves the claim. \square

3.1.4 The air cylinder

We consider the microstructure Σ_4 ; that is, an air cylinder with symmetry axis parallel to e_1 (see Fig. 2.2c). Combining the effective equations (2.12) with the index sets in Table 2.1, we obtain the effective system for this case:

$$\begin{cases} \hat{E} = 0 & \text{in } Q_M, \\ \hat{H}_2 = \hat{H}_3 = 0 & \text{in } Q_M. \end{cases} \quad (3.31a)$$

$$(3.31b)$$

As in the previous sections, we choose the following ansatz for the effective fields $\hat{E}, \hat{H}: G \rightarrow \mathbb{C}^3$,

$$\hat{E}(x) := (e^{-ik_0 x_1} + R e^{ik_0 x_1}) e_2 \quad \text{and} \quad \hat{H}(x) = -(e^{ik_0 x_1} - R e^{-ik_0 x_1}) e_3 \quad \text{for } x \in Q_R.$$

Equation (2.12b) determines the wave number and we find that $k_0 = \omega\sqrt{\varepsilon_0\mu_0}$. The effective electric field \hat{E} vanishes in the meta-material Q_M and hence, by (2.12a) and (3.31b), there is also no effective magnetic field in Q_M . So $\hat{E} = \hat{H} = 0$ in Q_M . Equation (2.12a) implies that the tangential trace of \hat{E} does not jump across the surface $\{x \in G: x_1 = 0\}$. Thus

$$R = -1.$$

As no field is transmitted through the meta-material Q_M , there is neither an electric nor an magnetic field in Q_L . In other words, $\hat{E} = \hat{H} = 0$ in Q_L . We have thus shown that

$$R = -1 \quad \text{and} \quad T = 0.$$

3.2 Vanishing limiting fields in high-contrast media, 2D-analysis

In this section, we perform an analysis of high-contrast media. Of the four geometries Σ_1 to Σ_4 , we study the two e_3 -invariant geometries: the metal cylinder Σ_1 and the metal plate Σ_3 , compare Fig. 2.2. We analyze the time-harmonic Maxwell equations (2.1) with the high-contrast permittivity ε_η of (2.10). We recall that the sequence of solutions $(E^\eta, H^\eta)_\eta$ is assumed to satisfy the $L^2(G)$ -bound (2.11). We are interested in the limit behaviour of $(E^\eta, H^\eta)_\eta$ as $\eta \rightarrow 0$.

When we consider perfect conductors, the effective equations (2.12) imply that some components of E^η or H^η converge weakly to 0 in the meta-material Q_M . For media with high-contrast, we do not have such a result (we recall that homogenization usually considers compactly contained geometries $\Sigma \subset Y$). In this section we ask for Σ_1 and Σ_3 : do the electric fields $(E^\eta)_\eta$ converge weakly in $L^2(Q_M; \mathbb{C}^3)$ to 0 as $\eta \rightarrow 0$? Is this weak convergence in fact a strong convergence? The same questions are considered for the magnetic fields $(H^\eta)_\eta$.

Let us point out that $E^\eta \cdot \mathbf{1}_{\Sigma_\eta} \rightarrow 0$ in $L^2(Q_M; \mathbb{C}^3)$. Indeed, the L^2 -estimate (2.11) can be improved to

$$\sup_{\eta > 0} \int_G \left(|\varepsilon_\eta| |E^\eta|^2 + |H^\eta|^2 \right) < \infty, \quad (3.32)$$

as was shown in [10, Section 3.1]. Thus

$$\frac{\varepsilon_1}{\eta^2} \int_{\Sigma_\eta} |E^\eta|^2 = \int_G |\varepsilon_\eta| |E^\eta|^2 \mathbf{1}_{\Sigma_\eta} \leq \int_G \left(|\varepsilon_\eta| |E^\eta|^2 + |H^\eta|^2 \right) \leq C. \quad (3.33)$$

So we have that $\|E^\eta \mathbf{1}_{\Sigma_\eta}\|_{L^2(G)}^2 \leq \eta^2 C$ which implies that $E^\eta \mathbf{1}_{\Sigma_\eta} \rightarrow 0$ in $L^2(Q_M; \mathbb{C}^3)$ as $\eta \rightarrow 0$.

We recall that the two geometries of interest are x_3 -independent. We therefore consider two different cases: In Section 3.2.1, we study electric fields E^η that are parallel to e_3 . In Section 3.2.2, we study magnetic fields H^η that are parallel to e_3 . By linearity of the equations, superpositions of these two cases provide the general behaviour of the material.

We will assume that the fields are x_3 -independent. This is a strong assumption, which can be justified for x_3 -independent incoming fields with a uniqueness property of solutions. In the rest of this section the fields $E^\eta(x)$ and $H^\eta(x)$ depend only on (x_1, x_2) .

Results for high-contrast media. In the E -parallel setting, the electric fields $(E^\eta)_\eta$ converge strongly to 0 in $L^2(Q_M; \mathbb{C}^3)$, the magnetic fields converge weakly to 0 in $L^2(Q_M; \mathbb{C}^3)$. On the other hand, when the magnetic fields H^η are parallel to e_3 , we can neither expect the electric fields nor the magnetic fields to converge weakly to 0 in $L^2(Q_M; \mathbb{C}^3)$.

3.2.1 Parallel electric field

We consider here the case of parallel electric fields, i.e., $E^\eta(x) := (0, 0, u^\eta(x))$ with $u^\eta = u^\eta(x_1, x_2)$. By abuse of notation, we will consider G also as a domain in \mathbb{R}^2 and write $(x_1, x_2) \in G$ when $(x_1, x_2, 0) \in G$; similarly for Σ_η . In this setting, the magnetic field H^η has no third component, $H^\eta(x) = (H_1^\eta(x_1, x_2), H_2^\eta(x_1, x_2), 0)$, and Maxwell's equations (2.1) reduce to the two-dimensional system

$$\begin{cases} -\nabla^\perp u^\eta = i\omega\mu_0(H_1^\eta, H_2^\eta) & \text{in } G, \\ \nabla^\perp \cdot (H_1^\eta, H_2^\eta) = -i\omega\varepsilon_0\varepsilon_\eta u^\eta & \text{in } G, \end{cases} \quad (3.34a)$$

$$(3.34b)$$

where we used the two-dimensional orthogonal gradient, $\nabla^\perp u := (-\partial_2 u, \partial_1 u)$, as well as the two-dimensional curl, $\nabla^\perp \cdot (H_1, H_2) := -\partial_2 H_1 + \partial_1 H_2$. The system (3.34) can equivalently be written as a scalar Helmholtz equation

$$-\Delta u^\eta = \omega^2 \varepsilon_0 \mu_0 \varepsilon_\eta u^\eta \quad \text{in } G \subset \mathbb{R}^2. \quad (3.35)$$

A solution of this Helmholtz equation provides the fields in the form $E^\eta = (0, 0, u^\eta)$ and $H^\eta = i(\omega\mu_0)^{-1}(\nabla^\perp u^\eta, 0)$.

Lemma 3.7 (Trivial limits for $E^\eta \parallel e_3$). *For $\eta > 0$ small, let $\Sigma_\eta \subset G \subset \mathbb{R}^2$ be a microscopic geometry that is given by Σ_1 or Σ_3 , and let the permittivity $\varepsilon_\eta: G \rightarrow \mathbb{C}$ be defined by (2.10). Let $E^\eta, H^\eta: G \rightarrow \mathbb{C}^3$ be solutions with $E^\eta(x) = (0, 0, u^\eta(x_1, x_2))$ that satisfy the estimate (2.11). Then*

$$E^\eta \rightarrow 0 \quad \text{and} \quad H^\eta \rightarrow 0 \quad \text{in } L^2(Q_M) \text{ as } \eta \rightarrow 0.$$

Proof. The L^2 -boundedness of E^η implies the L^2 -boundedness of u^η , and the L^2 -boundedness of H^η implies the L^2 -boundedness of ∇u^η . Therefore, the sequence $(u^\eta)_\eta$ is bounded in $H^1(G)$, and we find a limit function $u \in H^1(G)$ such that $u^\eta \rightharpoonup u$ in $H^1(G)$ and $u^\eta \rightarrow u$ in $L^2(G)$ as $\eta \rightarrow 0$.

We write

$$u^\eta \mathbb{1}_{Q_M} = u^\eta \mathbb{1}_{Q_M \setminus \Sigma_\eta} + u^\eta \mathbb{1}_{\Sigma_\eta}. \quad (3.36)$$

The left hand side converges strongly to $u \mathbb{1}_{Q_M}$. The first term on the right hand side of (3.36) is the product of a strongly $L^2(Q_M)$ -convergent sequence and a weakly $L^2(Q_M)$ -convergent sequence: $\mathbb{1}_{Q_M \setminus \Sigma_\eta} \rightharpoonup \alpha$ in $L^2(Q_M)$, where $\alpha \in (0, 1)$ is the volume fraction of $Y \setminus \Sigma$. We find that the first term on the right hand side converges in the sense of distributions to αu . The estimate (3.33) provides the strong convergence of the second term on the right hand side of (3.36) to zero. The distributional limit of (3.36) provides

$$u \mathbb{1}_{Q_M} = \alpha u \mathbb{1}_{Q_M} + 0, \quad (3.37)$$

and hence $u = 0$, since $\alpha \neq 0$. We have therefore found

$$E^\eta = (0, 0, u^\eta) \rightarrow 0 \quad \text{and} \quad H^\eta = (\nabla^\perp u^\eta, 0) \rightarrow 0 \quad \text{in } L^2(Q_M; \mathbb{C}^3) \text{ as } \eta \rightarrow 0,$$

which was the claim. \square

3.2.2 Parallel magnetic field

We now consider a magnetic field that is parallel to e_3 , $H^\eta(x) = (0, 0, u^\eta(x_1, x_2))$, with all quantities being x_3 -independent. This H -parallel case is the interesting case for homogenization and it has the potential to generate magnetically active materials. It was analyzed e.g. in [6, 7, 11, 23]. In this setting, Maxwell's equations (2.1) reduce to

$$\begin{cases} \nabla^\perp \cdot (E_1^\eta, E_2^\eta) = i\omega\mu_0 u^\eta & \text{in } G, \\ -\nabla^\perp u^\eta = -i\omega\varepsilon_0\varepsilon_\eta(E_1^\eta, E_2^\eta) & \text{in } G. \end{cases} \quad (3.38a)$$

$$(3.38b)$$

System (3.38) can equivalently be written as a scalar Helmholtz equation:

$$-\nabla \cdot \left(\frac{1}{\varepsilon_\eta} \nabla u^\eta \right) = \omega^2 \varepsilon_0 \mu_0 u^\eta \quad \text{in } G. \quad (3.39)$$

In (3.35), the high-contrast coefficient is outside the differential operator, which induces a trivial limit behaviour of solutions. Instead, (3.39) has the high-contrast coefficient inside the differential operator, which leads to a much richer behaviour of solutions.

The case $\Sigma = \Sigma_3$ is the metal plate (see Fig. 2.2b) that was studied in [11]. The result of [11] is the derivation of a limit system with nontrivial solutions. In particular, the weak limit of $(u^\eta)_\eta$ can be non-trivial. Similar results are available for metallic wires $\Sigma = \Sigma_1$ (see Fig. 2.2a); the results of [7] imply that also in this case the weak limit of $(u^\eta)_\eta$ can be non-trivial.

We therefore observe that the H -parallel case does not allow to conclude $H^\eta \rightarrow 0$ in $L^2(Q_M; \mathbb{C}^3)$. We note that $H^\eta = (0, 0, u^\eta) \not\rightarrow 0$ implies, by boundedness of the magnetic field and equation (3.38a), also $E^\eta \not\rightarrow 0$.

4 Finite element based multiscale approximation

In this section, we present numerical multiscale methods that are used to study Maxwell's equations in high-contrast media from a numerical point of view. We introduce the necessary notation for finite element discretisations and briefly discuss the utilized approaches. Based on these methods, numerical experiments illustrating the transmission properties of the microstructures are presented in Section 5.

4.1 Variational problem for the second order formulation

We study time-harmonic Maxwell's equations in their second-order formulation for the magnetic field H (2.3). The macroscopic domain \tilde{G} is assumed to be bounded and we impose impedance boundary conditions on the Lipschitz boundary $\partial\tilde{G}$ with the outer unit normal n :

$$\operatorname{curl} H \times n - ik_0(n \times H) \times n = g,$$

where $g \in L^2(\partial\tilde{G})$ with $g \cdot n = 0$ is given and $k_0 = \omega\sqrt{\varepsilon_0\mu_0}$ is the wavenumber. These boundary conditions can be interpreted as first-order approximation to the Silver-Müller radiation conditions (used for the full space \mathbb{R}^3); the data g are usually computed from an incident wave. For the material parameters, we choose $\mu = 1$ and ε_η as specified in 2.10. Multiplying with a test function and integrating by parts results in the following variational formulation: Find $H^\eta \in H_{\text{imp}}(\tilde{G})$ such that

$$\int_G \varepsilon_\eta^{-1} \operatorname{curl} H^\eta \cdot \operatorname{curl} \psi - k_0^2 H^\eta \cdot \psi \, dx - ik_0 \int_{\partial G} H_T^\eta \cdot \psi_T \, d\sigma = \int_{\partial G} g \cdot \psi_T \quad \forall \psi \in H_{\text{imp}}(\tilde{G}), \quad (4.1)$$

where $H_{\text{imp}}(\tilde{G}) := \{v \in L^2(\tilde{G}; \mathbb{C}^3) \mid \operatorname{curl} v \in L^2(\tilde{G}; \mathbb{C}^3), v_T \in L^2(\partial\tilde{G})\}$ and $v_T := v - (v \cdot n)n$ denotes the tangential component. Existence and uniqueness of the solution to this problem for fixed η is shown, for instance, in [31].

4.2 Traditional finite element discretisation

The standard finite element discretisation of (4.1) is a Galerkin procedure with a finite-dimensional approximation space $V_h \subset H_{\text{imp}}(\tilde{G})$ which consists of piecewise polynomial functions on a (tetrahedral) mesh of \tilde{G} . In detail, we denote by $\mathcal{T}_h = \{T_j \mid j \in J\}$ a partition of \tilde{G} into tetrahedra. We assume that \mathcal{T}_h is regular (i.e., no hanging nodes or edges occur), shape regular (i.e., the minimal angle stays bounded under mesh refinement), and that it resolves the partition into the meta-material Q_M and air $\tilde{G} \setminus \overline{Q}_M$. To allow for such a partition, we implicitly assume \tilde{G} and Q_M to be Lipschitz polyhedra. Otherwise, boundary approximations have to be used which only makes the following description more technical. We define the local mesh size $h_j := \operatorname{diam}(T_j)$ and the global mesh size $h := \max_{j \in J} h_j$. As conforming finite element space for $H_{\text{imp}}(\tilde{G})$, we use the lowest order edge elements introduced by Nédélec, i.e.,

$$V_h := \{v_h \in H_{\text{imp}}(\tilde{G}) \mid v_h|_K(x) = a + b \times x \text{ with } a, b \in \mathbb{C}^3, \forall K \in \mathcal{T}_h\}.$$

It is well known (see [31], for instance), that the finite element method with this test and trial space in (4.1) yields a well-posed discrete solution H_h . Furthermore, the following a priori error estimate holds

$$\|H^\eta - H_h\|_{H(\operatorname{curl})} \leq Ch(\|H^\eta\|_{H^1(\tilde{G})} + \|\operatorname{curl} H^\eta\|_{H^1(\tilde{G})}).$$

For the setting of (4.1) as discussed in this paper, however, two major problems arise. First, due to the discontinuities of the electric permittivity ε_η^{-1} the necessary regularity of H^η is not available, see [4, 18, 17]. Second, even in the case of sufficient regularity the right-hand side of the error estimate experiences a blow-up with $\|H^\eta\|_{H^1(\tilde{G})} + \|\operatorname{curl} H^\eta\|_{H^1(\tilde{G})} \rightarrow \infty$ for $\eta \rightarrow 0$. In other words, a typical solution of (4.1) is subject to (strong) oscillations in η such that its derivative does not remain bounded for the periodicity length tending to zero. As a consequence, the error estimate has a η -dependent right-hand side of the type $h\eta^{-1}$. Therefore, convergence of standard finite element discretisations is only to be expected in the asymptotic regime when $h \ll \eta$, i.e., the mesh has to resolve the oscillations in the PDE coefficients. As discussed in the introduction, this can become prohibitively expensive.

4.3 Heterogeneous Multiscale Method

As a remedy to these limitations of the standard finite element method, we consider a specific multiscale method. The idea is to extract macroscopic properties of the solution with η -independent complexity or computational effort, respectively. The basic idea directly comes up from the effective equation (2.14): Since this effective equation is independent of η , it can be discretised on a rather coarse mesh \mathcal{T}_h without the need to resolve the η -scale, i.e., we can have $h > \eta$. This results in an approximation of the homogenized solution \hat{H} , which contains important macroscopic information of H^η . However, for the discretisation of the homogenized equation, the effective material parameters $\hat{\varepsilon}$ and $\hat{\mu}$ need to be known, at least at the (macroscopic) quadrature points. This can be achieved by introducing another, again η -independent, mesh $\mathcal{T}_{h_Y} = \{S_l | l \in I\}$ of the unit cube Y with maximal mesh size $h_Y = \max_{l \in I} \operatorname{diam}(S_l)$. We assume that \mathcal{T}_{h_Y} is regular and shape regular and resolves the partition of Y into Σ and $Y \setminus \bar{\Sigma}$. Furthermore, \mathcal{T}_{h_Y} has to be periodic in the sense that it can be wrapped to a regular triangulation of the torus, i.e., no hanging nodes or edges over the periodic boundary. Note that h_Y denotes the mesh size of the triangulation of the unit cube. Thus, it is in no way related to η and can be of the same order as h . Based on this mesh, the cell problems occurring in the definition of $\hat{\varepsilon}$ and $\hat{\mu}$ can be discretised with standard (Lagrange and Nédélec) finite element spaces. For details we refer to [37]. All in all, we can now compute the homogenized solution \hat{H} of (2.14) as follows: 1. Compute discrete solutions of the cell problems (see [7] or [37]) using the mesh \mathcal{T}_{h_Y} and the associated standard finite element spaces. 2. Compute the effective parameters $\hat{\varepsilon}$ (or $\hat{\varepsilon}^{-1}$) and $\hat{\mu}$ approximatively with the discrete cell problems solutions. 3. Compute the discrete homogenized solution of (2.14) with the approximated effective coefficients and using the mesh \mathcal{T}_h with the associated finite element space V_h as introduced above.

This (naive) discretisation scheme for the effective equation (2.14) in fact can be interpreted as a specification of the Heterogeneous Multiscale Method (HMM) in the perfectly periodic case. The Finite Element Heterogeneous Multiscale Method, introduced by E and Enguist [19, 20], sets up a macroscopic sesquilinear form to compute the HMM solution H_h , which is an approximation of the homogenized solution \hat{H} . The macroscopic sesquilinear form is very similar to the effective sesquilinear form associated with the left-hand side of (2.14), but the effective material parameters are not computed a priori. Instead local variants of the cell problems are set up on η -scaled cubes $Y_j^\eta = \eta Y + x_j$ around macroscopic quadrature points x_j . We can still use the mesh \mathcal{T}_{h_Y} of the unit cube Y and transform it to a partition $\mathcal{T}_{h_Y}(Y_j^\eta)$ of the scaled unit cell Y_j^η . Similarly, also the finite element spaces associated with \mathcal{T}_{h_Y} can be transferred to spaces on $\mathcal{T}_{h_Y}(Y_j^\eta)$ using a suitable affine mapping. The finescale computations result in so called local reconstructions, which consist of the macroscopic basis functions and the corresponding (discrete) cell problem solutions. Averages (over Y_j^η) of these local reconstructions then enter the macroscopic sesquilinear form. A detailed definition of the HMM for Maxwell's equations in

high-contrast media is presented in [37], where also the connection to analytical homogenization as well as the possibility to treat more general than purely periodic problems are discussed. We only want to emphasize one important feature of the HMM in [37]: Apart from the (macroscopic) approximation H_h , discrete correctors $H_{h_Y,1}$, $H_{h_Y,2}$, and $H_{h_Y,3}$ can be determined from the discrete cell problems (in a second post-processing step). Via these correctors, we can define the zeroth order L^2 -approximation $H_{\text{HMM}}^0 := H_h + \nabla_y H_{h_Y,2}(\cdot, \frac{\cdot}{\eta}) + H_{h_Y,3}(\cdot, \frac{\cdot}{\eta})$, which corresponds to the first term of an asymptotic expansion and is used to approximate the true solution H^η . We again refer to [37] for details and note that it has been observed in several numerical examples that these correctors are a vital part of the HMM-approximation, see [25, 24, 33, 37]. We close by remarking that in Section 5 below, we extend the described HMM of [37] to general microstructures although the validity of the homogenized models in these cases is not shown so far, see the discussion in Sections 2.2 and 3.2.

5 Numerical study of transmission properties for high-contrast inclusions

In this section, we numerically study the transmission properties in the case of high-contrast for the three micro-geometries: the metal cylinder, the metal plate, and the air cylinder. Since the aim of this paper is a better understanding of the different microstructures and their effect, we focus on the qualitative behaviour rather than explicit convergence rates. The implementation was done with the module `dune-gdt` [30] of the DUNE software framework [3, 2].

Setting. We consider Maxwell's equations in the second-order formulation for the H -field (4.1) with a high-contrast medium as defined in (2.10). It remains to specify the macroscopic geometry, the boundary data g , the material parameter ε_1 , and the frequency. We use a slab-like macroscopic geometry similar to Section 2.1, but we truncate G also in the x_1 -direction to have a finite computational domain \tilde{G} , as described in the previous section. We choose $\tilde{G} = (0, 1)^3$ with the meta-material located in $Q_M = \{x \in G \mid 0.25 \leq x_1 \leq 0.75\}$. Note that Q_M is translated in x_1 -direction compared to Section 2.1, but this does not influence the qualitative results of the analysis. As in Section 2.1, we assume that an incident wave H_{inc} from the right travels along the x_1 -axis to the left, i.e., $H_{\text{inc}} = \exp(-ikx_1)p$ with a normalized polarization vector $p \perp \mathbf{e}_1$. This incident wave is used to compute the boundary data g as $g = \text{curl } H_{\text{inc}} \times n - ik_0 n \times (H_{\text{inc}} \times n)$. We choose the inverse permittivity as $\varepsilon_1^{-1} = 1.0 - 0.01i$ and note that ε_1 is only slightly dissipative. In all experiments, we choose the same wavenumber $k_0 = 12$ and the periodicity parameter $\eta = 1/8$.

As explained in the previous section, we want to use the Heterogeneous Multiscale Method to obtain good approximations with reasonable computational effort. We use the mesh sizes $h_y = h = \sqrt{3} \cdot 1/16$ and compute the macroscopic HMM-approximation H_h as well as the zeroth order approximation H_{HMM}^0 , which also utilizes information of the discrete correctors. To demonstrate the validity of the HMM, we use two different reference solutions. First, the homogenized reference solution \hat{H} is computed as solution to (2.14) on a mesh with size $h = \sqrt{3} \cdot 1/48$, where the effective material parameters are calculated approximatively using a discretisation of the unit cube with mesh size $h_Y = \sqrt{3} \cdot 1/24$. Second, the (true) reference solution H^η is computed as direct finite element discretisation of (4.1) on a fine grid with $h_{\text{ref}} = \sqrt{3} \cdot 1/64$.

Main results. Before we discuss the examples in detail, we present an overview of the results. The qualitative transmission properties of the meta-material are in good agreement with the theory of Section 3.1, although the numerical examples consider high-contrast media instead of perfect conductors. The predictions and the corresponding numerical examples are summarized in Table 5.1. In contrast to perfect conductors, the high-contrast medium leads to rather high

Table 5.1: Summary of analytical predictions of the transmission properties and references to numerical results. The first row provides the geometry. The second row indicates possible transmission polarizations (of H) according to the theory of perfect conductors of Section 3.1. The third row indicates the possibility of transmission based on Section 3.2: We mention cases in which we cannot derive weak convergence to 0. An entry “-” indicates that no analytical result can be applied. The last row provides the reference to the visualization of the numerical calculation for high-contrast media.

geometry	metal cylinder Σ_1	metal cylinder Σ_2	metal plate Σ_3	air cyl. Σ_4
transmission (PC)	\mathbf{e}_3 -polarized	\mathbf{e}_2 and \mathbf{e}_3 -polarized	\mathbf{e}_3 -polarized	no
nontriv. limit (HC)	\mathbf{e}_3 -polarized	-	\mathbf{e}_3 -polarized	-
numerical example	Fig. 5.1	Fig. 5.3	Fig. 5.4	Fig. 5.5

intensities and amplitudes of the H -field inside the inclusions Σ_η . Depending on the chosen wavenumber, Mie-resonances inside the inclusions can occur for high-contrast media; see Section 3.2 and [7, 37]. Our numerical experiments also show that the HMM yields (qualitatively) good approximations, although the validity of the underlying effective models is not proved for the studied geometries.

5.1 Metal cuboids $\tilde{\Sigma}_1$ and $\tilde{\Sigma}_2$

Instead of metal cylinders with circular base we study metal cuboids with square base, so that we do not have to deal with boundary approximations in our numerical method. We choose $\tilde{\Sigma}_1 = (0.25, 0.75)^2 \times (0, 1)$ and $\tilde{\Sigma}_2 = (0, 1) \times (0.25, 0.75)^2$. Note that this choice influences the value of γ , but not the other results of Sections 3.1.1 and 3.1.2. Due to the symmetry of the microstructure, the effective material parameters are diagonal matrices with $a_{11} = a_{22}$, but with a different value a_{33} ; see the analytical computations in Section 3.1.1. Up to numerical errors, we obtain the same structure for the computed approximative effective parameters.

Comparing the homogenized reference solution \tilde{H} for \mathbf{e}_2 - and \mathbf{e}_3 -polarized incoming waves for $\tilde{\Sigma}_1$ in Fig. 5.1, we observe that the \mathbf{e}_3 -polarized wave is transmitted almost undisturbed through the meta-material. For the \mathbf{e}_2 -polarization, however, the field intensity in $Q_L := \{x \in G: x \leq 0.25\}$ is very low, corresponding to small transmission factors. This matches the analytical predictions of Section 3.1.1, which yields transmission only for \mathbf{e}_3 -parallel H -fields. The same effect is predicted for high-contrast media by the analysis of Section 3.2.1.

The HMM can reproduce the behaviour of the homogenized and of the heterogeneous solution. For the comparison, we only consider the \mathbf{e}_3 -polarized incoming wave in Fig. 5.2 and compare the zeroth order approximation H_{HMM}^0 (right) to the (true) reference solution H^η (left). Errors are still visible, but the qualitative agreement is good, even for the coarse mesh size of $h = h_Y = \sqrt{3} \cdot 1/16$ chosen for the HMM. In particular, the rather cheaply computable zeroth order approximation H_{HMM}^0 can capture most of the important features of the *true* solution, even for inclusions of high-contrast. This clearly underlines the potential of the HMM. Moreover, Fig. 5.2 underlines the specific behaviour of H^η in the inclusions $\tilde{\Sigma}_\eta$ for high-contrast media. As analyzed in Section 3.2.2, H^η cannot be expected to vanish in the inclusions in the limit $\eta \rightarrow 0$ due to possible resonances; see [6]. We observe rather high field intensities in the inclusions; see also [37] for a slightly different inclusion geometry.

We also study the rotated metal cuboid $\tilde{\Sigma}_2$. In correspondence to the analysis of Section 3.1.2, we observe transmission for an \mathbf{e}_3 -polarized incident wave; see Fig. 5.3. Note that the

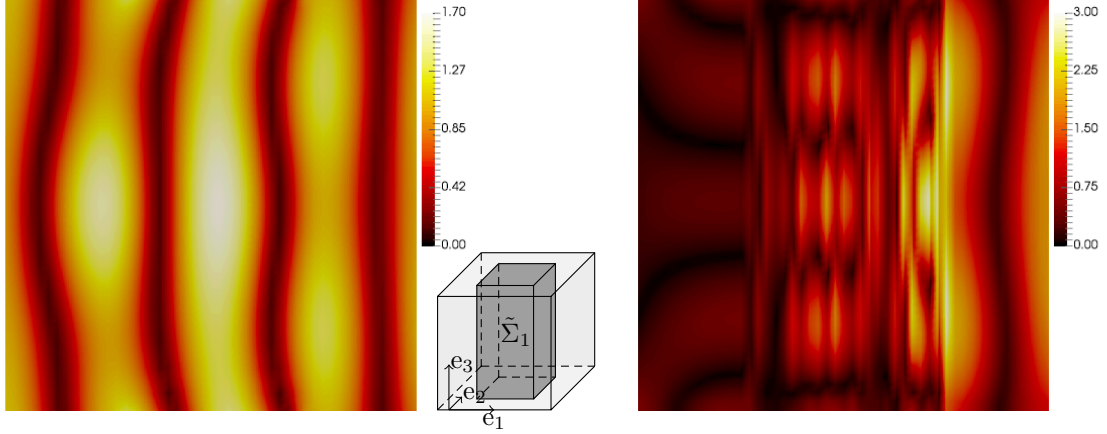


Figure 5.1: Metal cuboid $\tilde{\Sigma}_1$, the magnitude of $\text{Re}(\hat{H})$ is plotted. Left: The H -field is e_3 -polarized and the plot shows values in the plane $x_3 = 0.5$. The analysis of both, (PC) and (HC) yields: transmission is possible. Right: The H -field is e_2 -polarized and the plot shows values in the plane $x_2 = 0.545$. Since the H -field is not parallel to e_3 , the analysis of (PC) and (HC) predicts that no transmission is possible. Inlet in the middle: Microstructure in the unit cube.

homogenized reference solution looks different to $\tilde{\Sigma}_1$ because of the rotation of the geometry, which is also reflected in the different structure and values of the reflection and transmission coefficients. The (true) reference solution in Fig. 5.4 shows the high field intensities in the metal cuboids induced by the high-contrast permittivity.

5.2 Metal plate Σ_3

As in Section 3.1.3, we choose a metal plate perpendicular to e_2 of width 0.5, i.e. $\Sigma_3 = (0, 1) \times (0.25, 0.75) \times (0, 1)$. Discretising the cell problems with mesh size $h_Y = \sqrt{3} \cdot 1/24$, we obtain—up to numerical errors—the effective material parameters as diagonal matrices with

$$\begin{aligned} \widehat{\varepsilon}^{-1} &\approx \text{diag}(10^{-4}, 0.5, 10^{-4}), \\ \text{Re } \hat{\mu} &\approx \text{diag}(0.228303, -0.044672, 0.228303). \end{aligned}$$

Although we consider high-contrast media, this correspond astonishingly well to the analytical results for perfect conductors of Section 3.1.3: The structure of the matrices agrees and the non-zero value of $\widehat{\varepsilon}^{-1} = |Y \setminus \bar{\Sigma}|$ is as expected from the theory of perfect conductors. Due to the contributions of the inclusions, the values of μ_{hom} are different from the case of perfect conductors.

Section 3.1.3 shows that, for perfect conductors, only an H -field polarized in e_3 -direction can be transmitted through the meta-material. Our numerical experiments allow a similar observation for high-contrast media in Fig. 5.4: The homogenized reference solution only shows a non-negligible intensity in the domain $Q_L = \{x \in G | x_3 \leq 0.25\}$ left of the scatterer if the incident wave is polarized in e_3 direction. Note that we have some reflections from the boundary in Fig. 5.4 since we do not use perfectly matched layers as boundary conditions. The observed transmission properties for high-contrast media are in accordance with the theory in Section 3.2: For an e_3 -polarized H -field as in the left figure, we cannot expect a (weak) convergence to zero.

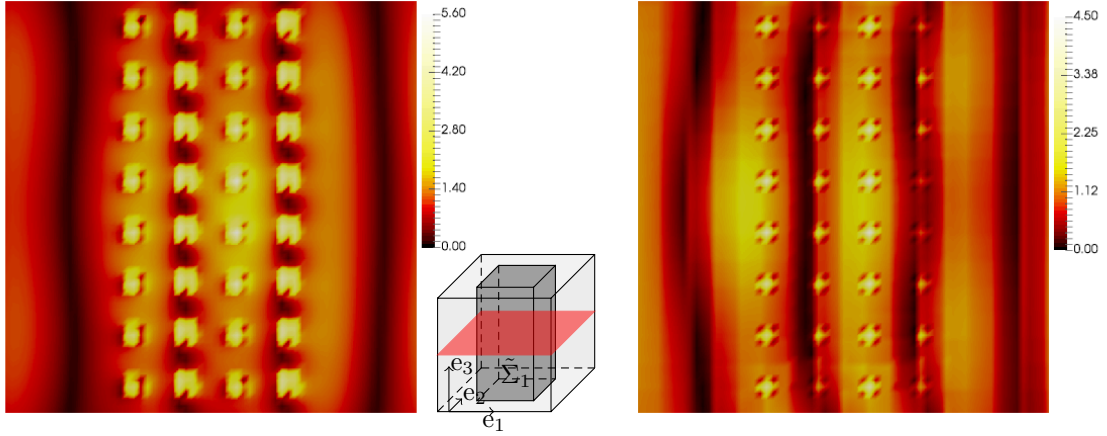


Figure 5.2: Test of numerical schemes for the metal cuboid $\tilde{\Sigma}_1$. We consider an e_3 -polarized incoming H -field and plot the solution in the plane $x_3 = 0.5$; the colors indicate the magnitude of the reference solution $\text{Re}(H^\eta)$ (left) and the zeroth order approximation $\text{Re}(H_{\text{HMM}}^0)$ (right). Inlet in the center: Microstructure in the unit cube with visualization plane in red.

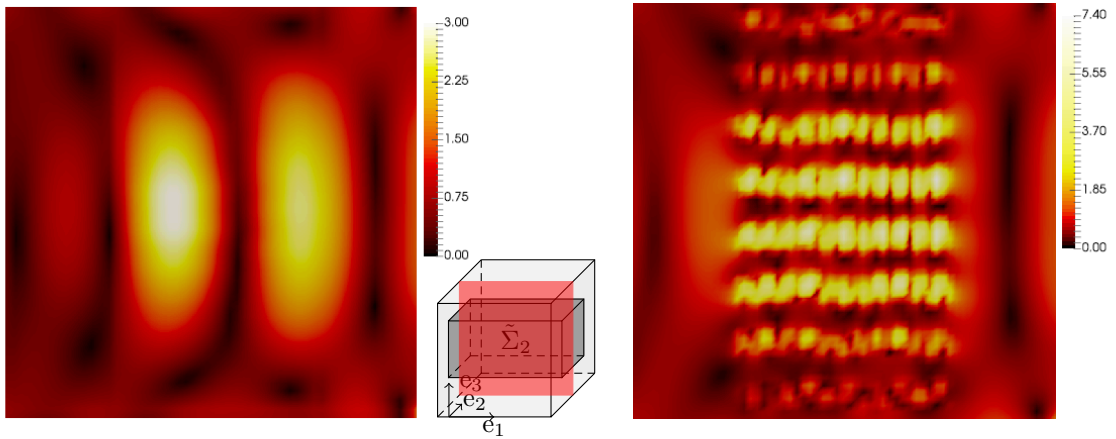


Figure 5.3: Metal cuboid $\tilde{\Sigma}_2$. We study an e_3 -polarized incident H -field and plot the magnitude of $\text{Re}(\hat{H})$ (left) and $\text{Re}(H^\eta)$ (right) in the plane $x_2 = 0.545$. The analysis (PC) predicts transmission in this case, the analysis (HC) does not exclude transmission. Middle: Microstructure in the unit cube with visualization plane in red.

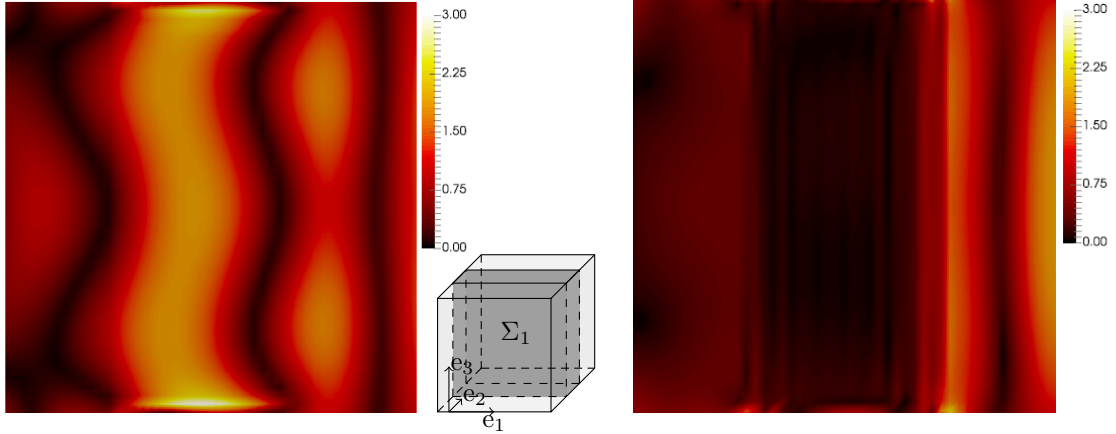


Figure 5.4: Metal plate Σ_3 . The colors indicate the magnitude of $\text{Re}(\hat{H})$ in the plane $x_3 = 0.5$. Left: The H -field is e_3 -polarized. The analysis (PC) predicts transmission, the analysis (HC) cannot exclude transmission. Right: The H -field is e_2 -polarized. The analysis (PC and HC) predicts that no transmission is possible.

This corresponds to the observed non-trivial transmission. By contrast, in the right figure, the H -field is e_2 -polarized and no transmission can be observed. This corresponds to the analysis of Section 3.2.1, which shows that H^η converges to zero, weakly in $L^2(Q_M)$.

5.3 Air cuboid $\tilde{\Sigma}_4$

As with the metal cylinder, we equip the air cylinder of Section 3.1.4 with a square base in order to have a geometry-fitting mesh. To be precise, we define the microstructure $\tilde{\Sigma}_4 = (0, 1)^3 \setminus ((0, 1) \times (0.25, 0.75)^2)$. The effective permittivity $\widehat{\varepsilon}^{-1}$ vanishes almost identically for this setting; numerically we obtain only entries of order 10^{-5} for a discretisation of the corresponding cell problem with mesh size $h_Y = \sqrt{3} \cdot 1/24$. As discussed in Section 3.1.4, no transmission through this meta-material is expected for the high conductors. We observe the same for high-contrast media in Fig. 5.5: The (true) reference solution (almost) vanishes in the left part Q_L in all situations. Here, we only depict e_3 -polarized incident waves, once for $\tilde{\Sigma}_4$ as described and once for the rotated air cuboid with main axis in e_3 -direction (this is the setting of Section 3.1.1 with interchanged roles of metal and air). Note that inside the microstructure, high intensities and amplitudes of the H^η -field occur due to resonances in the high-contrast medium.

Conclusion

We analyzed the transmission properties of meta-materials consisting of perfect conductors or high-contrast materials. Depending on the geometry of the microstructure, certain entries in the effective material parameters vanish, which induces that also certain components of the solution vanish. This influences the transmission properties of the material. Transmission is possible only for certain polarizations of the incoming wave. For perfect conductors, we derived closed formulas for the reflection and transmission coefficients. Using the Heterogeneous Multiscale Method, the homogenized solution as well as some features of the exact solution can be approximated on rather coarse meshes and, in particular, with a cost that is independent of the periodicity length. Our

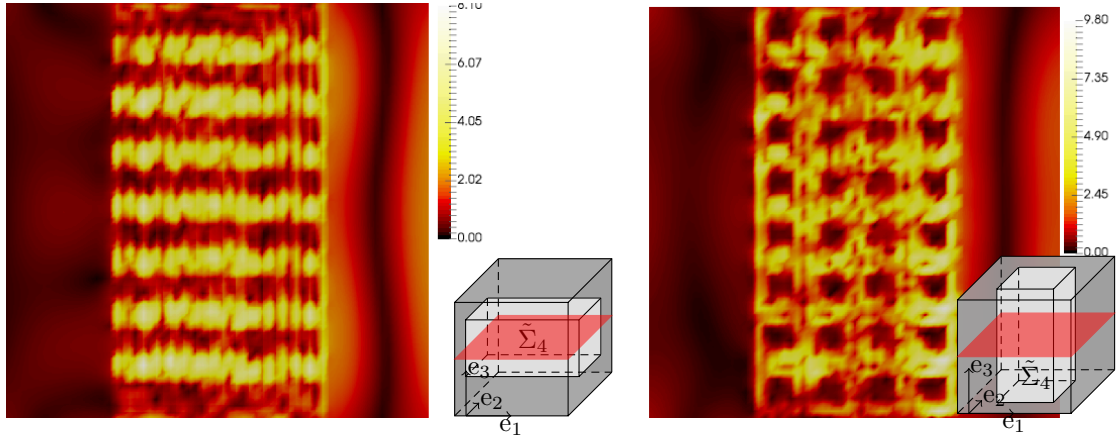


Figure 5.5: Metal block with holes. Left: The structure $\tilde{\Sigma}_4$, we plot the magnitude of $\text{Re}(H^n)$ in the plane $x_3 = 0.545$ for e_3 -polarized incoming H -field. The analysis (PC) predicts no transmission, the analysis (HC) cannot exclude transmission. Right: A geometry in which the cylinders $\tilde{\Sigma}_4$ are rotated in e_3 -direction. We plot the magnitude of $\text{Re}(H^n)$ in the plane $x_3 = 0.5$ for e_3 -polarized incoming H -field. Small pictures show the microstructures in the unit cube and the visualization planes in red.

numerical experiments of three representative geometries with high-contrast materials confirm the theoretical predictions of their transmission properties.

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